DETERMINISTIC CRAMÉR-RAO BOUND FOR SYMMETRIC PARAFAC MODEL WITH APPLICATION TO BLIND SPATIAL SIGNATURE ESTIMATION

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ABSTRACT

The symmetric PARAllel FACtor analysis (PARAFAC) model has found numerous applications in array signal processing and communications. In this paper, we derive the deterministic Cramér-Rao Bound (CRB) for the symmetric RARAFAC model and illustrate the obtained results using an example with spatial signature estimation in sensor arrays.

1. INTRODUCTION AND DATA MODEL

A family of blind array processing algorithms including ESPRIT-like method [1]-[2], Second Order Blind Identification (SOBI) algorithm [3]-[4] and blind spatial signature estimation method based on time-varying user power loading [5] exploit the models which essentially share the same structure called a *symmetric* PARAFAC model.

The CRB analysis for the methods based on the symmetric PARAFAC model is of great interest. In this paper, we derive such CRB in a closed form and illustrate the obtained results using an example with spatial signature estimation in sensor arrays.

Let an array of K sensors receive the signals from M narrowband sources. The $K\times 1$ snapshot vector of antenna array outputs can be written as

$$y(n) = As(n) + v(n) \tag{1}$$

where $A = [a_1, \ldots, a_M]$ is the $K \times M$ complex matrix of the user spatial signatures, $a_m = [a_{1,m}, \ldots, a_{K,m}]^T$ is the $K \times 1$ complex spatial signature of the mth user, $s(n) = [s_1(n), \ldots, s_M(n)]^T$ is the $M \times 1$ complex vector of the user waveforms, $v(n) = [v_1(n), \ldots, v_K(n)]^T$ is the $K \times 1$ vector of additive spatially and temporally white complex Gaussian noise, and $(\cdot)^T$ denotes the transpose. Assuming that there is a block of N snapshots available, the model (1) can be written as

$$Y = AS + V \tag{2}$$

where $Y = [y(1), \ldots, y(N)]$ is the $K \times N$ array data matrix, $S = [s(1), \ldots, s(N)]$ is the $M \times N$ user waveform matrix, and $V = [v(1), \ldots, v(N)]$ is the $K \times N$ sensor noise matrix.

Assuming that the user signals are uncorrelated with each other and sensor noise, the array covariance matrix of the received signals can be written as

$$\mathbf{R} = \mathbb{E}\{\mathbf{y}(n)\mathbf{y}^{H}(n)\} = \mathbf{A}\mathbf{Q}\mathbf{A}^{H} + \sigma^{2}\mathbf{I}$$
(3)

where $Q = E\{s(n)s^H(n)\}$ is the diagonal covariance matrix of the signal waveforms, σ^2 is the sensor noise variance, I is the identity matrix, and $(\cdot)^H$ denotes the Hermitian transpose.

2. SYMMETRIC PARAFAC MODEL

Often, it is required to estimate the matrix A in (2) based on the observations Y only. In the multiple user case, this is not possible to do with only one known covariance matrix (3) because the matrix A can be estimated from R only up to an arbitrary unknown unitary matrix. To provide a unique estimate of A, several covariance matrices have to be used, see [1]-[5].

In this paper, following the approach of [5] with artificial user power loading, we assume that a set of covariance matrices is obtained by dividing uniformly the whole data block of N snapshots into P sub-blocks, each of $N_s = \lfloor \frac{N}{P} \rfloor$ snapshots, where $\lfloor x \rfloor$ denotes the largest integer less than x. The transmitted power of each user is assumed to be fixed within each particular sub-block while is changed from one sub-block to another. Using such power loading scheme, we obtain that the received snapshots within any pth sub-block correspond to the following covariance matrix

$$R(p) = AQ(p)A^{H} + \sigma^{2}I$$
 (4)

where Q(p) is the diagonal covariance matrix of the user waveforms in the pth sub-block and $p = 1, \ldots, P$.

In practice, the noise power can be estimated and then subtracted from the covariance matrix (4). Let us stack the P matrices $R(p) - \sigma^2 I$, $p = 1, \ldots, P$ together to form a three-way array \underline{R} . This three-way array has a symmetry dictated by the symmetry of the matrices $R(p) - \sigma^2 I$. The (i, l, p)th element of such an array can be written as

$$r_{i,l,p} = [\underline{R}]_{i,l,p} = \sum_{m=1}^{M} a_{i,m} \nu_m(p) a_{l,m}^*$$
 (5)

where $\nu_m(p) = [Q(p)]_{m,m}$ is the power of the mth user in the pth sub-block and $(\cdot)^*$ denotes the complex conjugate. Defining the $P \times M$ matrix P as

$$\mathbf{P} = \begin{bmatrix} \nu_1(1) & \dots & \nu_M(1) \\ \vdots & \ddots & \vdots \\ \nu_1(P) & \dots & \nu_M(P) \end{bmatrix}$$
 (6)

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we have that $Q(p) = \mathcal{D}_p\{P\}$ for all p = 1, ..., P where $\mathcal{D}_p\{\cdot\}$ is the operator that makes a diagonal matrix by selecting the pth row and putting it on the main diagonal while putting zeros elsewhere.

3. DETERMINISTIC CRAMÉR-RAO BOUND

The model (1) for the nth sample of the pth sub-block can be rewritten as

$$y(p,n) = AQ^{1/2}(p)\tilde{s}(n) + v(n),$$

 $n = (p-1)N_s + 1, ..., pN_s$ (7)

where $\tilde{s}(n) = [\tilde{s}_1(n), \dots, \tilde{s}_M(n)]^T = Q^{-1/2}(p)s(n)$ is the vector of normalized signal waveforms and the normalization is done so that all waveforms have unit powers.

Hence, the observations in the pth sub-block satisfy the following model

$$\mathbf{y}(p,n) \sim \mathcal{CN}(\boldsymbol{\mu}(p,n), \sigma^2 \mathbf{I})$$
 (8)

where

$$\mu(p,n) = AQ^{1/2}(p)\tilde{s}(n), \quad n = (p-1)N_s + 1, \dots, pN_s$$
 (9)

The unknown parameters of the model (7) are all the entries of A, the diagonal elements of Q(p) $(p=1,\ldots,P)$ and the noise power σ^2 . Note, however, that the latter parameter is decoupled with the other parameters in the Fisher Information Matrix (FIM) [6]. Therefore, without loss of generality, σ^2 can be excluded from the vector of unknown parameters.

A delicate point regarding the CRB is the inherent permutation and scale ambiguity. To derive a meaningful CRB, we assume that the first row of \boldsymbol{A} is normalized to $[1,\ldots,1]_{1\times M}$ (this removes the scaling ambiguity), and the first row of \boldsymbol{P} is known and consists of distinct elements (which resolves the permutation ambiguity). Then, the $2(K-1)M\times 1$ real vector of the unknown parameters is given by

$$\alpha = [\alpha_2^T, \dots, \alpha_K^T]^T \tag{10}$$

where $\alpha_k = [\operatorname{Re}\{\tilde{a}_k\}^T, \operatorname{Im}\{\tilde{a}_k\}^T]^T$ and $\tilde{a}_k = [a_{k,1}, \dots, a_{k,M}]^T$. The $(P-1)M \times 1$ vector of nuisance parameters can be expressed as

$$\boldsymbol{\zeta} = [\tilde{\boldsymbol{p}}(2), \dots, \tilde{\boldsymbol{p}}(P)]^T \tag{11}$$

where $\tilde{p}(p)$ is the pth row of the matrix P.

Using (10) and (11), the $(2(K-1)M + (P-1)M) \times 1$ real vector of unknown parameters can be written as

$$\boldsymbol{\theta} = [\boldsymbol{\alpha}^T, \ \boldsymbol{\zeta}^T]^T \tag{12}$$

THEOREM: The $(2(K-1)M+(P-1)M)\times(2(K-1)M+(P-1)M)$ FIM is given by

$$\begin{bmatrix} J_{\alpha_2,\alpha_2} & 0 & & & \\ & \ddots & & & & \\ \hline 0 & J_{\alpha,\tilde{p}(2)} & \dots & J_{\alpha,\tilde{p}(P)} \\ \hline & J_{\alpha,\tilde{p}(2)}^T & & J_{\tilde{p}(2),\tilde{p}(2)} & 0 \\ & \vdots & & & \ddots \\ & J_{\alpha,\tilde{p}(P)}^T & & 0 & & J_{\tilde{p}(P),\tilde{p}(P)} \end{bmatrix}$$

where

$$J_{\alpha_{2},\alpha_{2}} = \cdots = J_{\alpha_{K},\alpha_{K}}$$

$$= \frac{2}{\sigma^{2}} \begin{bmatrix} \operatorname{Re}\{\Upsilon^{H}\Upsilon\} & \operatorname{Im}\{\Upsilon^{H}\Upsilon\} \\ \operatorname{Im}\{\Upsilon^{H}\Upsilon\} & \operatorname{Re}\{\Upsilon^{H}\Upsilon\} \end{bmatrix}$$
(13)

$$J_{\tilde{p}(p),\tilde{p}(p)} = \frac{2}{\sigma^2} \operatorname{Re}\{(G(p))^H G(p)\}$$
(14)

$$\boldsymbol{J}_{\boldsymbol{\alpha},\bar{\boldsymbol{p}}(p)} = \frac{2}{\sigma^2} (\boldsymbol{I}_{K-1} \otimes \tilde{\boldsymbol{F}}(p)) \tilde{\boldsymbol{H}}(p) \tag{15}$$

$$\Upsilon = \begin{bmatrix} f_1(1) & \dots & f_M(1) \\ \vdots & \ddots & \vdots \\ f_1(P) & \dots & f_M(P) \end{bmatrix}$$
 (16)

$$G(p) = \begin{bmatrix} h_{1,1}(p) & \dots & h_{1,M}(p) \\ \vdots & \ddots & \vdots \\ h_{K,1}(p) & \dots & h_{K,M}(p) \end{bmatrix}$$
(17)

$$\tilde{\boldsymbol{F}}(p) = \begin{bmatrix} \operatorname{Re}\{\boldsymbol{F}^{H}(p)\} & \operatorname{Im}\{\boldsymbol{F}^{H}(p)\} \\ \operatorname{Im}\{\boldsymbol{F}^{H}(p)\} & \operatorname{Re}\{\boldsymbol{F}^{H}(p)\} \end{bmatrix}$$
(18)

$$F(p) = [f_1(p), \dots, f_M(p)]$$
 (19)

$$\tilde{\boldsymbol{H}}(p) = \left[\tilde{\boldsymbol{H}}_{2}^{T}(p), \dots, \tilde{\boldsymbol{H}}_{K}^{T}(p)\right]^{T}$$
(20)

$$\tilde{\boldsymbol{H}}_{k}(p) = \begin{bmatrix} \operatorname{Re}\{\boldsymbol{H}_{k}(p)\} \\ \operatorname{Im}\{\boldsymbol{H}_{k}(p)\} \end{bmatrix}$$
 (21)

$$H_k(p) = [h_{k,1}(p), \dots, h_{k,M}(p)]$$
 (22)

$$\mathbf{f}_m(p) = \begin{bmatrix} \sqrt{\nu_m(p)} \bar{s}_m((p-1)N_s + 1), \end{bmatrix}$$

$$\dots, \sqrt{\nu_m(p)}\tilde{s}_m(pN_s)\Big]^T$$
 (23)

$$\boldsymbol{h}_{k,m}(p) = \left[\frac{a_{k,m} \tilde{s}_m((p-1)N_s + 1)}{2\sqrt{\nu_m(p)}}, \dots, \frac{a_{k,m} \tilde{s}_m(pN_s)}{2\sqrt{\nu_m(p)}} \right]^T$$
(24)

and \otimes denotes the Kronecker matrix product.

The $(K-1)M \times (K-1)M$ spatial signature-related block of the CRB matrix is given in the closed form as

$$CRB_{\alpha,\alpha} = \left[\boldsymbol{J}_{\alpha,\alpha} - \frac{2}{\sigma^2} \sum_{p=2}^{P} (\boldsymbol{I}_{K-1} \otimes \tilde{\boldsymbol{F}}(p)) \tilde{\boldsymbol{H}}(p) \right] \times \left[Re\{\boldsymbol{G}^H(p)\boldsymbol{G}(p)\} \right]^{-1} \tilde{\boldsymbol{H}}^H(p) (\boldsymbol{I}_{K-1} \otimes \tilde{\boldsymbol{F}}(p))^{H} \right]^{-1} (25)$$

where the upper-left block of the FIM can be expressed as

$$J_{\alpha,\alpha} = \frac{2}{\sigma^2} I_{K-1} \otimes \begin{bmatrix} \operatorname{Re}\{\Upsilon^H \Upsilon\} & \operatorname{Im}\{\Upsilon^H \Upsilon\} \\ \operatorname{Im}\{\Upsilon^H \Upsilon\} & \operatorname{Re}\{\Upsilon^H \Upsilon\} \end{bmatrix}$$
(26)

PROOF: The (l, k)th element of the FIM is given by [6]

$$FIM_{l,k} = \frac{2}{\sigma^2}$$

$$\times \sum_{p=1}^{P} \sum_{n=(p-1)N_c+1}^{pN_s} \operatorname{Re}\left(\frac{\partial \boldsymbol{\mu}^H(p,n)}{\partial \theta_l} \frac{\partial \boldsymbol{\mu}(p,n)}{\partial \theta_k}\right)$$
(27)

Using (9) along with (27), we have

$$\frac{\partial \mu(p,n)}{\partial \operatorname{Re}\{a_{k,m}\}} = \sqrt{\nu_m(p)} \tilde{s}_m(n) e_k \tag{28}$$

$$\frac{\partial \boldsymbol{\mu}(p,n)}{\partial \operatorname{Im}\{a_{k,m}\}} = j\sqrt{\nu_m(p)}\bar{s}_m(n)e_k \tag{29}$$

$$\frac{\partial \mu(p,n)}{\partial \nu_m(p)} = \left[\frac{a_{1,m} \tilde{s}_m(n)}{2\sqrt{\nu_m(p)}}, \dots, \frac{a_{K,m} \tilde{s}_m(n)}{2\sqrt{\nu_m(p)}} \right]^T \tag{30}$$

where e_k is the vector containing one in the kth position and zeros elsewhere.

Using (28) and (29) along with (27) we obtain that

$$J_{\text{Re}\{a_{k,m}\},\text{Re}\{a_{k,l}\}} = J_{\text{Im}\{a_{k,m}\},\text{Im}\{a_{k,l}\}}$$

$$= \frac{2}{\sigma^{2}} \sum_{p=1}^{P} \sum_{n=(p-1)N_{s}+1}^{pN_{s}} \text{Re} \left\{ \sqrt{\nu_{m}(p)\nu_{l}(p)} \tilde{s}_{m}^{*}(n) \tilde{s}_{l}(n) \right\}$$

$$= \frac{2}{\sigma^{2}} \text{Re} \{ \xi_{m}^{H} \xi_{l} \}$$
(31)

where $\boldsymbol{\xi}_m = [\boldsymbol{f}_m^T(1), \dots, \boldsymbol{f}_m^T(P)]^T$. Similarly,

$$J_{\text{Im}\{a_{k,m}\},\text{Re}\{a_{k,l}\}} = J_{\text{Re}\{a_{k,m}\},\text{Im}\{a_{k,l}\}}$$

$$= \frac{2}{\sigma^2} \text{Im}\{\xi_m^H \xi_l\} \qquad (32)$$

Therefore,

$$J_{\operatorname{Re}\{\alpha_{k}\},\operatorname{Re}\{\alpha_{k}\}} = J_{\operatorname{Im}\{\alpha_{k}\},\operatorname{Im}\{\alpha_{k}\}}$$

$$= \frac{2}{\sigma^{2}} \begin{bmatrix} \operatorname{Re}\{\xi_{1}^{H}\xi_{1}\} & \dots & \operatorname{Re}\{\xi_{1}^{H}\xi_{M}\} \\ \vdots & \ddots & \vdots \\ \operatorname{Re}\{\xi_{M}^{H}\xi_{1}\} & \dots & \operatorname{Re}\{\xi_{M}^{H}\xi_{M}\} \end{bmatrix}$$

$$= \frac{2}{\sigma^{2}} \operatorname{Re}\{\Upsilon^{H}\Upsilon\}$$
(33)

and

$$J_{\operatorname{Im}\{\alpha_{k}\},\operatorname{Re}\{\alpha_{k}\}} = J_{\operatorname{Re}\{\alpha_{k}\},\operatorname{Im}\{\alpha_{k}\}}$$

$$= \frac{2}{\sigma^{2}} \begin{bmatrix} \operatorname{Im}\{\xi_{1}^{H}\xi_{1}\} & \dots & \operatorname{Im}\{\xi_{1}^{H}\xi_{M}\} \\ \vdots & \ddots & \vdots \\ \operatorname{Im}\{\xi_{M}^{H}\xi_{1}\} & \dots & \operatorname{Im}\{\xi_{M}^{H}\xi_{M}\} \end{bmatrix}$$

$$= \frac{2}{\sigma^{2}} \operatorname{Im}\{\Upsilon^{H}\Upsilon\}$$
(34)

Using (33) and (34), we obtain (13). Note that the right-hand side of (13) does not depend on the index k. Hence,

$$J_{\alpha,\alpha} = \begin{bmatrix} J_{\alpha_{2},\alpha_{2}} & 0 \\ & \ddots \\ 0 & J_{\alpha_{K},\alpha_{K}} \end{bmatrix}$$
$$= \frac{2}{\sigma^{2}} I_{K-1} \otimes \begin{bmatrix} \operatorname{Re}\{\Upsilon^{H}\Upsilon\} & \operatorname{Im}\{\Upsilon^{H}\Upsilon\} \\ \operatorname{Im}\{\Upsilon^{H}\Upsilon\} & \operatorname{Re}\{\Upsilon^{H}\Upsilon\} \end{bmatrix} (35)$$

Next, using (30) along with (27) we can write for $p=2,\ldots,P$ and $m,l=1,\ldots,M$

$$\begin{aligned}
& \left[J_{\tilde{p}(p),\tilde{p}(p)} \right]_{m,l} = \frac{2}{\sigma^2} \\
\times & \sum_{n=(p-1)N_s+1}^{pN_s} \sum_{k=1}^K \operatorname{Re} \left\{ \frac{\left(a_{k,m} \tilde{s}_m(n) \right)^*}{2\sqrt{\nu_m(p)}} \frac{a_{k,l} \tilde{s}_l(n)}{2\sqrt{\nu_l(p)}} \right\} \\
&= \frac{2}{\sigma^2} \operatorname{Re} \{ c_m^H(p) c_l(p) \}
\end{aligned} (36)$$

where $c_m(p) = [h_{1,m}^T(p), \dots, h_{K,m}^T(p)]^T$. Stacking all M^2 elements given by (36) in one matrix we have for $p = 2, \dots, P$

$$J_{\hat{p}(p),\hat{p}(p)} = \frac{2}{\sigma^2}$$

$$\times \begin{bmatrix} \operatorname{Re}\{\boldsymbol{c}_1^H(p)\boldsymbol{c}_1(p)\} & \dots & \operatorname{Re}\{\boldsymbol{c}_1^H(p)\boldsymbol{c}_M(p)\} \\ \vdots & \ddots & \vdots \\ \operatorname{Re}\{\boldsymbol{c}_M^H(p)\boldsymbol{c}_1(p)\} & \dots & \operatorname{Re}\{\boldsymbol{c}_M^H(p)\boldsymbol{c}_M(p)\} \end{bmatrix}$$

$$= \frac{2}{\sigma^2}\operatorname{Re}\{\boldsymbol{G}^H(p)\boldsymbol{G}(p)\}$$
(37)

Finally, using (28), (29), and (30) along with (27) we can write for $p=2,\ldots P;\,k=2,\ldots,K,$ and $m,l=1,\ldots,M$

$$\left[J_{\operatorname{Re}\{a_{k}\},\tilde{p}(p)}\right]_{m,l} = \frac{2}{\sigma^{2}}$$

$$\times \sum_{n=(p-1)N_{s}+1}^{pN_{s}} \operatorname{Re}\left\{\frac{1}{2} \frac{\sqrt{\nu_{m}(p)}}{\sqrt{\nu_{l}(p)}} \tilde{s}_{m}^{*}(n) a_{k,l} \tilde{s}_{l}(n)\right\}$$

$$= \frac{2}{\sigma^{2}} \operatorname{Re}\{f_{m}^{H}(p) h_{k,l}(p)\} \tag{38}$$

$$\begin{aligned}
& \left[J_{\text{Im}\{a_{k}\},\tilde{p}(p)}\right]_{m,l} = \frac{2}{\sigma^{2}} \\
& \times \sum_{n=(p-1)N_{s}+1}^{pN_{s}} \text{Re} \left\{ j \frac{1}{2} \frac{\sqrt{\nu_{m}(p)}}{\sqrt{\nu_{l}(p)}} \tilde{s}_{m}^{\star}(n) a_{k,l} \tilde{s}_{l}(n) \right\} \\
& = \frac{2}{\sigma^{2}} \text{Im} \{f_{m}^{H}(p) h_{k,l}(p)\}
\end{aligned} (39)$$

Collecting all $(K-1)M^2$ elements given by (38) and $(K-1)M^2$ elements given by (39) in one matrix, we obtain for $p=2,\ldots,P$

$$J_{\alpha,\hat{p}(p)} = \frac{2}{\sigma^{2}} \begin{bmatrix} \begin{bmatrix} \operatorname{Re}\{\boldsymbol{F}^{H}(p)\boldsymbol{H}_{2}(p)\} \\ \operatorname{Im}\{\boldsymbol{F}^{H}(p)\boldsymbol{H}_{2}(p)\} \end{bmatrix} \\ \vdots \\ \begin{bmatrix} \operatorname{Re}\{\boldsymbol{F}^{H}(p)\boldsymbol{H}_{K}(p)\} \\ \operatorname{Im}\{\boldsymbol{F}^{H}(p)\boldsymbol{H}_{K}(p)\} \end{bmatrix} \end{bmatrix}$$
(40)

Observing that

$$\begin{bmatrix} \operatorname{Re}\{\boldsymbol{F}^{H}(p)\boldsymbol{H}_{k}(p)\} \\ \operatorname{Im}\{\boldsymbol{F}^{H}(p)\boldsymbol{H}_{k}(p)\} \end{bmatrix} = \tilde{\boldsymbol{F}}(p)\tilde{\boldsymbol{H}}_{k}(p)$$
(41)

we can further simplify (40) to

$$\boldsymbol{J}_{\boldsymbol{\alpha},\tilde{\boldsymbol{p}}(p)} = \frac{2}{\sigma^2} \left(\boldsymbol{I}_{K-1} \otimes \tilde{\boldsymbol{F}}(p) \right) \tilde{\boldsymbol{H}}(p) \tag{42}$$

Also, note that

$$\boldsymbol{J}_{\boldsymbol{\alpha},\tilde{\boldsymbol{p}}(p)}^{T} = \boldsymbol{J}_{\tilde{\boldsymbol{p}}(p),\boldsymbol{\alpha}} \tag{43}$$

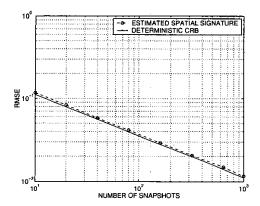


Fig. 1. CRB and RMSE versus N.

Using (35), (37), (42) and (43) we obtain the expressions (13)-(24).

Computing the CRB for θ requires the inverse of the $(2(K-1)M+(P-1)M)\times(2(K-1)M+(P-1)M)$ FIM matrix. Our objective is to obtain the CRB associated with the vector parameter α only, avoiding the inverse of the full FIM matrix. Exploiting the fact that the lower-right sub-block of the FIM is a block-diagonal matrix and using the partitioned matrix inversion lemma (see [6], p. 572), after some algebra we obtain (25)-(26) and the proof is complete.

4. SIMULATIONS

In order to test the derived CRB we consider a simple example with spatial signature estimation of a single user and assume that the BPSK signal impinges on the linear array of 4 sensors and unknown geometry from $\theta=50^\circ$ relative to the broadside direction. It is well known that in the single-user case, a single covariance matrix is sufficient to guarantee the uniqueness of the spatial signature estimate which is given by the principal eigenvector of the sample covariance matrix \hat{R} .

We compare the Root-Mean-Square Error (RMSE) performance of such a principal eigenvector-based estimator with the derived CRB. The RMSE is computed as

$$RMSE = \sqrt{\frac{1}{LK} \sum_{l=1}^{L} \|\hat{\boldsymbol{a}}(l) - \boldsymbol{a}\|_F^2}$$

where L=100 is the number of independent simulation runs and $\hat{a}(l)$ is the estimate of a obtained in the lth run. Note that the scaling ambiguity is eliminated by normalizing $\hat{a}(l)$ with respect to the first (reference) sensor. The CRB is computed as

$$CRB = \sqrt{\frac{1}{K-1}} Tr\{CRB_{\alpha,\alpha}\}$$

Figure 1 displays the RMSE and the CRB versus the number of snapshots N for the Signal-to-Noise Ratio (SNR) equal to 10 dB. Figure 2 shows the same quantities versus the SNR for N=100

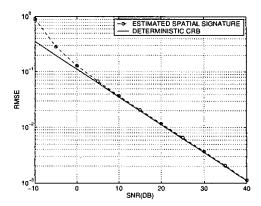


Fig. 2. CRB and RMSE versus SNR.

It can be seen that the principal eigenvector-based spatial signature estimator approaches CRB at high SNR. This validates our CRB analysis.

5. CONCLUSIONS

The closed-form expressions for the deterministic CRB for the symmetric PARAFAC model have been derived. The simulation example with blind spatial signature estimation illustrates and validates our CRB analysis.

6. REFERENCES

- [1] D. Astèly, A. L. Swindlehurst and B. Ottersten, "Spatial signature estimation for uniform linear arrays with unknown receiver gains and phases," *IEEE Trans. Signal Processing*, vol. 47, pp. 2128-2138, Aug. 1999.
- [2] M. K. Tsatsanis and C. Kweon, "Blind source separation of non-stationary sources using second-order statistics," in *Proc. 32nd Asilomar Conf. Signals, Systems and Computers*, Pacific Grove, CA, Nov. 1998, vol. 2, pp. 1574 -1578.
- [3] A. Belouchrani, K. Abed-Meraim, J-F. Cardoso and E. Moulines, "A blind source separation technique using second-order statistics," *IEEE Trans. on Signal Processing*, vol. 45, No.2, pp. 434-444, Feb. 1997.
- [4] A. Yeredor, "Non-orthogonal joint diagonalization in the least-squares sense with application in blind source separation," *IEEE Trans. Signal Processing*, vol. 50, pp. 1545-1553, July 2002.
- [5] Y. Rong, S. A. Vorobyov, A. B. Gershman and N. D. Sidiropoulos, "Blind spatial signature estimation using time-varying user power loading and parallel factor analysis," *Proc. IEEE VTC'03 Fall*, Orlando, USA, Oct. 2003 (also submitted to *IEEE Trans. Signal Processing*).
- [6] S. M. Kay, Fundamentals of Statistical Signal Processing: Estimation Theory, Englewood Cliffs, NJ: Prentice-Hall, 1993.