

# DETERMINISTIC CRAMÉR-RAO BOUND FOR SYMMETRIC PARAFAC MODEL WITH APPLICATION TO BLIND SPATIAL SIGNATURE ESTIMATION

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## ABSTRACT

The symmetric PARAllel FACTor analysis (PARAFAC) model has found numerous applications in array signal processing and communications. In this paper, we derive the deterministic Cramér-Rao Bound (CRB) for the symmetric PARAFAC model and illustrate the obtained results using an example with spatial signature estimation in sensor arrays.

## 1. INTRODUCTION AND DATA MODEL

A family of blind array processing algorithms including ESPRIT-like method [1]-[2], Second Order Blind Identification (SOBI) algorithm [3]-[4] and blind spatial signature estimation method based on time-varying user power loading [5] exploit the models which essentially share the same structure called a *symmetric* PARAFAC model.

The CRB analysis for the methods based on the symmetric PARAFAC model is of great interest. In this paper, we derive such CRB in a closed form and illustrate the obtained results using an example with spatial signature estimation in sensor arrays.

Let an array of  $K$  sensors receive the signals from  $M$  narrow-band sources. The  $K \times 1$  snapshot vector of antenna array outputs can be written as

$$\mathbf{y}(n) = \mathbf{A}\mathbf{s}(n) + \mathbf{v}(n) \quad (1)$$

where  $\mathbf{A} = [\mathbf{a}_1, \dots, \mathbf{a}_M]$  is the  $K \times M$  complex matrix of the user spatial signatures,  $\mathbf{a}_m = [a_{1,m}, \dots, a_{K,m}]^T$  is the  $K \times 1$  complex spatial signature of the  $m$ th user,  $\mathbf{s}(n) = [s_1(n), \dots, s_M(n)]^T$  is the  $M \times 1$  complex vector of the user waveforms,  $\mathbf{v}(n) = [v_1(n), \dots, v_K(n)]^T$  is the  $K \times 1$  vector of additive spatially and temporally white complex Gaussian noise, and  $(\cdot)^T$  denotes the transpose. Assuming that there is a block of  $N$  snapshots available, the model (1) can be written as

$$\mathbf{Y} = \mathbf{A}\mathbf{S} + \mathbf{V} \quad (2)$$

where  $\mathbf{Y} = [\mathbf{y}(1), \dots, \mathbf{y}(N)]$  is the  $K \times N$  array data matrix,  $\mathbf{S} = [\mathbf{s}(1), \dots, \mathbf{s}(N)]$  is the  $M \times N$  user waveform matrix, and  $\mathbf{V} = [\mathbf{v}(1), \dots, \mathbf{v}(N)]$  is the  $K \times N$  sensor noise matrix.

Assuming that the user signals are uncorrelated with each other and sensor noise, the array covariance matrix of the received signals can be written as

$$\mathbf{R} = \mathbb{E}\{\mathbf{y}(n)\mathbf{y}^H(n)\} = \mathbf{A}\mathbf{Q}\mathbf{A}^H + \sigma^2\mathbf{I} \quad (3)$$

where  $\mathbf{Q} = \mathbb{E}\{\mathbf{s}(n)\mathbf{s}^H(n)\}$  is the diagonal covariance matrix of the signal waveforms,  $\sigma^2$  is the sensor noise variance,  $\mathbf{I}$  is the identity matrix, and  $(\cdot)^H$  denotes the Hermitian transpose.

## 2. SYMMETRIC PARAFAC MODEL

Often, it is required to estimate the matrix  $\mathbf{A}$  in (2) based on the observations  $\mathbf{Y}$  only. In the multiple user case, this is not possible to do with only one known covariance matrix (3) because the matrix  $\mathbf{A}$  can be estimated from  $\mathbf{R}$  only up to an arbitrary unknown unitary matrix. To provide a unique estimate of  $\mathbf{A}$ , several covariance matrices have to be used, see [1]-[5].

In this paper, following the approach of [5] with artificial user power loading, we assume that a set of covariance matrices is obtained by dividing uniformly the whole data block of  $N$  snapshots into  $P$  sub-blocks, each of  $N_p = \lfloor \frac{N}{P} \rfloor$  snapshots, where  $\lfloor x \rfloor$  denotes the largest integer less than  $x$ . The transmitted power of each user is assumed to be fixed within each particular sub-block while is changed from one sub-block to another. Using such power loading scheme, we obtain that the received snapshots within any  $p$ th sub-block correspond to the following covariance matrix

$$\mathbf{R}(p) = \mathbf{A}\mathbf{Q}(p)\mathbf{A}^H + \sigma^2\mathbf{I} \quad (4)$$

where  $\mathbf{Q}(p)$  is the diagonal covariance matrix of the user waveforms in the  $p$ th sub-block and  $p = 1, \dots, P$ .

In practice, the noise power can be estimated and then subtracted from the covariance matrix (4). Let us stack the  $P$  matrices  $\mathbf{R}(p) - \sigma^2\mathbf{I}$ ,  $p = 1, \dots, P$  together to form a three-way array  $\underline{\mathbf{R}}$ . This three-way array has a symmetry dictated by the symmetry of the matrices  $\mathbf{R}(p) - \sigma^2\mathbf{I}$ . The  $(i, l, p)$ th element of such an array can be written as

$$r_{i,l,p} = [\underline{\mathbf{R}}]_{i,l,p} = \sum_{m=1}^M a_{i,m}\nu_m(p)a_{l,m}^* \quad (5)$$

where  $\nu_m(p) = [\mathbf{Q}(p)]_{m,m}$  is the power of the  $m$ th user in the  $p$ th sub-block and  $(\cdot)^*$  denotes the complex conjugate. Defining the  $P \times M$  matrix  $\mathbf{P}$  as

$$\mathbf{P} = \begin{bmatrix} \nu_1(1) & \dots & \nu_M(1) \\ \vdots & \ddots & \vdots \\ \nu_1(P) & \dots & \nu_M(P) \end{bmatrix} \quad (6)$$



Using (9) along with (27), we have

$$\frac{\partial \mu(p, n)}{\partial \text{Re}\{a_{k,m}\}} = \sqrt{\nu_m(p)} \bar{s}_m(n) e_k \quad (28)$$

$$\frac{\partial \mu(p, n)}{\partial \text{Im}\{a_{k,m}\}} = j \sqrt{\nu_m(p)} \bar{s}_m(n) e_k \quad (29)$$

$$\frac{\partial \mu(p, n)}{\partial \nu_m(p)} = \left[ \frac{a_{1,m} \bar{s}_m(n)}{2\sqrt{\nu_m(p)}}, \dots, \frac{a_{K,m} \bar{s}_m(n)}{2\sqrt{\nu_m(p)}} \right]^T \quad (30)$$

where  $e_k$  is the vector containing one in the  $k$ th position and zeros elsewhere.

Using (28) and (29) along with (27) we obtain that

$$\begin{aligned} \mathbf{J}_{\text{Re}\{a_{k,m}\}, \text{Re}\{a_{k,l}\}} &= \mathbf{J}_{\text{Im}\{a_{k,m}\}, \text{Im}\{a_{k,l}\}} \\ &= \frac{2}{\sigma^2} \sum_{p=1}^P \sum_{n=(p-1)N_s+1}^{pN_s} \text{Re} \left\{ \sqrt{\nu_m(p)} \nu_l(p) \bar{s}_m^*(n) \bar{s}_l(n) \right\} \\ &= \frac{2}{\sigma^2} \text{Re}\{\boldsymbol{\xi}_m^H \boldsymbol{\xi}_l\} \end{aligned} \quad (31)$$

where  $\boldsymbol{\xi}_m = [\mathbf{f}_m^T(1), \dots, \mathbf{f}_m^T(P)]^T$ .

Similarly,

$$\begin{aligned} \mathbf{J}_{\text{Im}\{a_{k,m}\}, \text{Re}\{a_{k,l}\}} &= \mathbf{J}_{\text{Re}\{a_{k,m}\}, \text{Im}\{a_{k,l}\}} \\ &= \frac{2}{\sigma^2} \text{Im}\{\boldsymbol{\xi}_m^H \boldsymbol{\xi}_l\} \end{aligned} \quad (32)$$

Therefore,

$$\begin{aligned} \mathbf{J}_{\text{Re}\{\boldsymbol{\alpha}_k\}, \text{Re}\{\boldsymbol{\alpha}_k\}} &= \mathbf{J}_{\text{Im}\{\boldsymbol{\alpha}_k\}, \text{Im}\{\boldsymbol{\alpha}_k\}} \\ &= \frac{2}{\sigma^2} \begin{bmatrix} \text{Re}\{\boldsymbol{\xi}_1^H \boldsymbol{\xi}_1\} & \dots & \text{Re}\{\boldsymbol{\xi}_1^H \boldsymbol{\xi}_M\} \\ \vdots & \ddots & \vdots \\ \text{Re}\{\boldsymbol{\xi}_M^H \boldsymbol{\xi}_1\} & \dots & \text{Re}\{\boldsymbol{\xi}_M^H \boldsymbol{\xi}_M\} \end{bmatrix} \\ &= \frac{2}{\sigma^2} \text{Re}\{\boldsymbol{\Upsilon}^H \boldsymbol{\Upsilon}\} \end{aligned} \quad (33)$$

and

$$\begin{aligned} \mathbf{J}_{\text{Im}\{\boldsymbol{\alpha}_k\}, \text{Re}\{\boldsymbol{\alpha}_k\}} &= \mathbf{J}_{\text{Re}\{\boldsymbol{\alpha}_k\}, \text{Im}\{\boldsymbol{\alpha}_k\}} \\ &= \frac{2}{\sigma^2} \begin{bmatrix} \text{Im}\{\boldsymbol{\xi}_1^H \boldsymbol{\xi}_1\} & \dots & \text{Im}\{\boldsymbol{\xi}_1^H \boldsymbol{\xi}_M\} \\ \vdots & \ddots & \vdots \\ \text{Im}\{\boldsymbol{\xi}_M^H \boldsymbol{\xi}_1\} & \dots & \text{Im}\{\boldsymbol{\xi}_M^H \boldsymbol{\xi}_M\} \end{bmatrix} \\ &= \frac{2}{\sigma^2} \text{Im}\{\boldsymbol{\Upsilon}^H \boldsymbol{\Upsilon}\} \end{aligned} \quad (34)$$

Using (33) and (34), we obtain (13). Note that the right-hand side of (13) does not depend on the index  $k$ . Hence,

$$\begin{aligned} \mathbf{J}_{\boldsymbol{\alpha}, \boldsymbol{\alpha}} &= \begin{bmatrix} \mathbf{J}_{\boldsymbol{\alpha}_2, \boldsymbol{\alpha}_2} & & 0 \\ & \ddots & \\ 0 & & \mathbf{J}_{\boldsymbol{\alpha}_K, \boldsymbol{\alpha}_K} \end{bmatrix} \\ &= \frac{2}{\sigma^2} \mathbf{I}_{K-1} \otimes \begin{bmatrix} \text{Re}\{\boldsymbol{\Upsilon}^H \boldsymbol{\Upsilon}\} & \text{Im}\{\boldsymbol{\Upsilon}^H \boldsymbol{\Upsilon}\} \\ \text{Im}\{\boldsymbol{\Upsilon}^H \boldsymbol{\Upsilon}\} & \text{Re}\{\boldsymbol{\Upsilon}^H \boldsymbol{\Upsilon}\} \end{bmatrix} \end{aligned} \quad (35)$$

Next, using (30) along with (27) we can write for  $p = 2, \dots, P$  and  $m, l = 1, \dots, M$

$$\begin{aligned} [\mathbf{J}_{\hat{\mathbf{p}}(p), \hat{\mathbf{p}}(p)}]_{m,l} &= \frac{2}{\sigma^2} \\ &\times \sum_{n=(p-1)N_s+1}^{pN_s} \sum_{k=1}^K \text{Re} \left\{ \frac{(a_{k,m} \bar{s}_m(n))^* a_{k,l} \bar{s}_l(n)}{2\sqrt{\nu_m(p)} 2\sqrt{\nu_l(p)}} \right\} \\ &= \frac{2}{\sigma^2} \text{Re}\{\mathbf{c}_m^H(p) \mathbf{c}_l(p)\} \end{aligned} \quad (36)$$

where  $\mathbf{c}_m(p) = [\mathbf{h}_{1,m}^T(p), \dots, \mathbf{h}_{K,m}^T(p)]^T$ . Stacking all  $M^2$  elements given by (36) in one matrix we have for  $p = 2, \dots, P$

$$\begin{aligned} \mathbf{J}_{\hat{\mathbf{p}}(p), \hat{\mathbf{p}}(p)} &= \frac{2}{\sigma^2} \\ &\times \begin{bmatrix} \text{Re}\{\mathbf{c}_1^H(p) \mathbf{c}_1(p)\} & \dots & \text{Re}\{\mathbf{c}_1^H(p) \mathbf{c}_M(p)\} \\ \vdots & \ddots & \vdots \\ \text{Re}\{\mathbf{c}_M^H(p) \mathbf{c}_1(p)\} & \dots & \text{Re}\{\mathbf{c}_M^H(p) \mathbf{c}_M(p)\} \end{bmatrix} \\ &= \frac{2}{\sigma^2} \text{Re}\{\mathbf{G}^H(p) \mathbf{G}(p)\} \end{aligned} \quad (37)$$

Finally, using (28), (29), and (30) along with (27) we can write for  $p = 2, \dots, P$ ;  $k = 2, \dots, K$ , and  $m, l = 1, \dots, M$

$$\begin{aligned} [\mathbf{J}_{\text{Re}\{\boldsymbol{\alpha}_k\}, \hat{\mathbf{p}}(p)}]_{m,l} &= \frac{2}{\sigma^2} \\ &\times \sum_{n=(p-1)N_s+1}^{pN_s} \text{Re} \left\{ \frac{1}{2} \frac{\sqrt{\nu_m(p)}}{\sqrt{\nu_l(p)}} \bar{s}_m^*(n) a_{k,l} \bar{s}_l(n) \right\} \\ &= \frac{2}{\sigma^2} \text{Re}\{\mathbf{f}_m^H(p) \mathbf{h}_{k,l}(p)\} \end{aligned} \quad (38)$$

$$\begin{aligned} [\mathbf{J}_{\text{Im}\{\boldsymbol{\alpha}_k\}, \hat{\mathbf{p}}(p)}]_{m,l} &= \frac{2}{\sigma^2} \\ &\times \sum_{n=(p-1)N_s+1}^{pN_s} \text{Re} \left\{ j \frac{1}{2} \frac{\sqrt{\nu_m(p)}}{\sqrt{\nu_l(p)}} \bar{s}_m^*(n) a_{k,l} \bar{s}_l(n) \right\} \\ &= \frac{2}{\sigma^2} \text{Im}\{\mathbf{f}_m^H(p) \mathbf{h}_{k,l}(p)\} \end{aligned} \quad (39)$$

Collecting all  $(K-1)M^2$  elements given by (38) and  $(K-1)M^2$  elements given by (39) in one matrix, we obtain for  $p = 2, \dots, P$

$$\mathbf{J}_{\boldsymbol{\alpha}, \hat{\mathbf{p}}(p)} = \frac{2}{\sigma^2} \begin{bmatrix} \begin{bmatrix} \text{Re}\{\mathbf{F}^H(p) \mathbf{H}_2(p)\} \\ \text{Im}\{\mathbf{F}^H(p) \mathbf{H}_2(p)\} \end{bmatrix} \\ \vdots \\ \begin{bmatrix} \text{Re}\{\mathbf{F}^H(p) \mathbf{H}_K(p)\} \\ \text{Im}\{\mathbf{F}^H(p) \mathbf{H}_K(p)\} \end{bmatrix} \end{bmatrix} \quad (40)$$

Observing that

$$\begin{bmatrix} \text{Re}\{\mathbf{F}^H(p) \mathbf{H}_k(p)\} \\ \text{Im}\{\mathbf{F}^H(p) \mathbf{H}_k(p)\} \end{bmatrix} = \tilde{\mathbf{F}}(p) \tilde{\mathbf{H}}_k(p) \quad (41)$$

we can further simplify (40) to

$$\mathbf{J}_{\boldsymbol{\alpha}, \hat{\mathbf{p}}(p)} = \frac{2}{\sigma^2} (\mathbf{I}_{K-1} \otimes \tilde{\mathbf{F}}(p)) \tilde{\mathbf{H}}(p) \quad (42)$$

Also, note that

$$\mathbf{J}_{\boldsymbol{\alpha}, \hat{\mathbf{p}}(p)}^T = \mathbf{J}_{\hat{\mathbf{p}}(p), \boldsymbol{\alpha}} \quad (43)$$

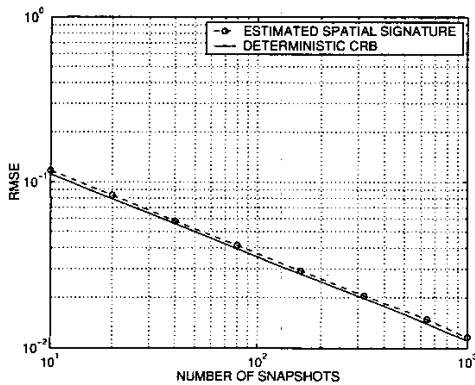


Fig. 1. CRB and RMSE versus  $N$ .

Using (35), (37), (42) and (43) we obtain the expressions (13)-(24).

Computing the CRB for  $\theta$  requires the inverse of the  $(2(K-1)M + (P-1)M) \times (2(K-1)M + (P-1)M)$  FIM matrix. Our objective is to obtain the CRB associated with the vector parameter  $\alpha$  only, avoiding the inverse of the full FIM matrix. Exploiting the fact that the lower-right sub-block of the FIM is a block-diagonal matrix and using the partitioned matrix inversion lemma (see [6], p. 572), after some algebra we obtain (25)-(26) and the proof is complete.

#### 4. SIMULATIONS

In order to test the derived CRB we consider a simple example with spatial signature estimation of a single user and assume that the BPSK signal impinges on the linear array of 4 sensors and unknown geometry from  $\theta = 50^\circ$  relative to the broadside direction. It is well known that in the single-user case, a single covariance matrix is sufficient to guarantee the uniqueness of the spatial signature estimate which is given by the principal eigenvector of the sample covariance matrix  $\hat{R}$ .

We compare the Root-Mean-Square Error (RMSE) performance of such a principal eigenvector-based estimator with the derived CRB. The RMSE is computed as

$$\text{RMSE} = \sqrt{\frac{1}{LK} \sum_{l=1}^L \|\hat{a}(l) - \alpha\|_2^2}$$

where  $L = 100$  is the number of independent simulation runs and  $\hat{a}(l)$  is the estimate of  $\alpha$  obtained in the  $l$ th run. Note that the scaling ambiguity is eliminated by normalizing  $\hat{a}(l)$  with respect to the first (reference) sensor. The CRB is computed as

$$\text{CRB} = \sqrt{\frac{1}{K} \text{Tr}\{\text{CRB}_{\alpha, \alpha}\}}$$

Figure 1 displays the RMSE and the CRB versus the number of snapshots  $N$  for the Signal-to-Noise Ratio (SNR) equal to 10 dB. Figure 2 shows the same quantities versus the SNR for  $N = 100$ .

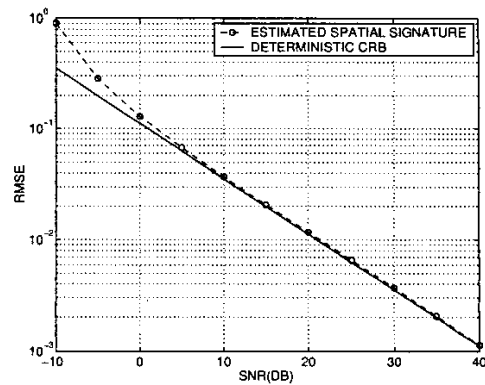


Fig. 2. CRB and RMSE versus SNR.

It can be seen that the principal eigenvector-based spatial signature estimator approaches CRB at high SNR. This validates our CRB analysis.

#### 5. CONCLUSIONS

The closed-form expressions for the deterministic CRB for the symmetric PARAFAC model have been derived. The simulation example with blind spatial signature estimation illustrates and validates our CRB analysis.

#### 6. REFERENCES

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